



RBE 2004

ระบบอัตโนมัติ (Automatic System)

สาขาวิศวกรรมหุ่นยนต์

คณะวิศวกรรมศาสตร์และเทคโนโลยีอุตสาหกรรม

มหาวิทยาลัยราชภัฏสวนสุนันทา

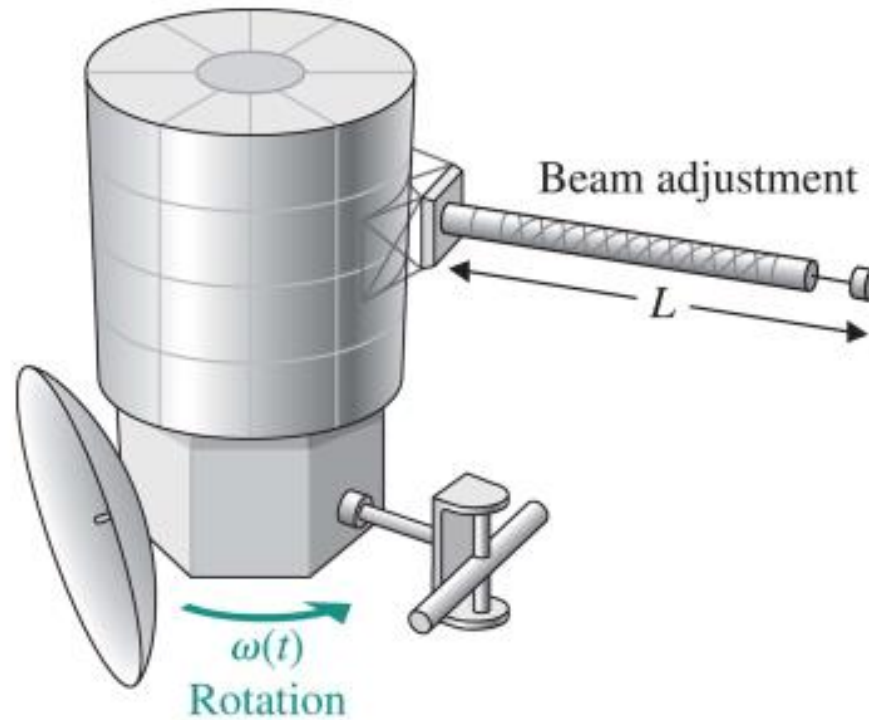
Chapter 2 Mathematic Model of System

Lecture 4 Effect of pole locations

- Understand the concept of transient response
- Observe the influence of the pole location in the system response
- Find the response of a system for a certain input
- Computer Simulation (Matlab/Simulink)

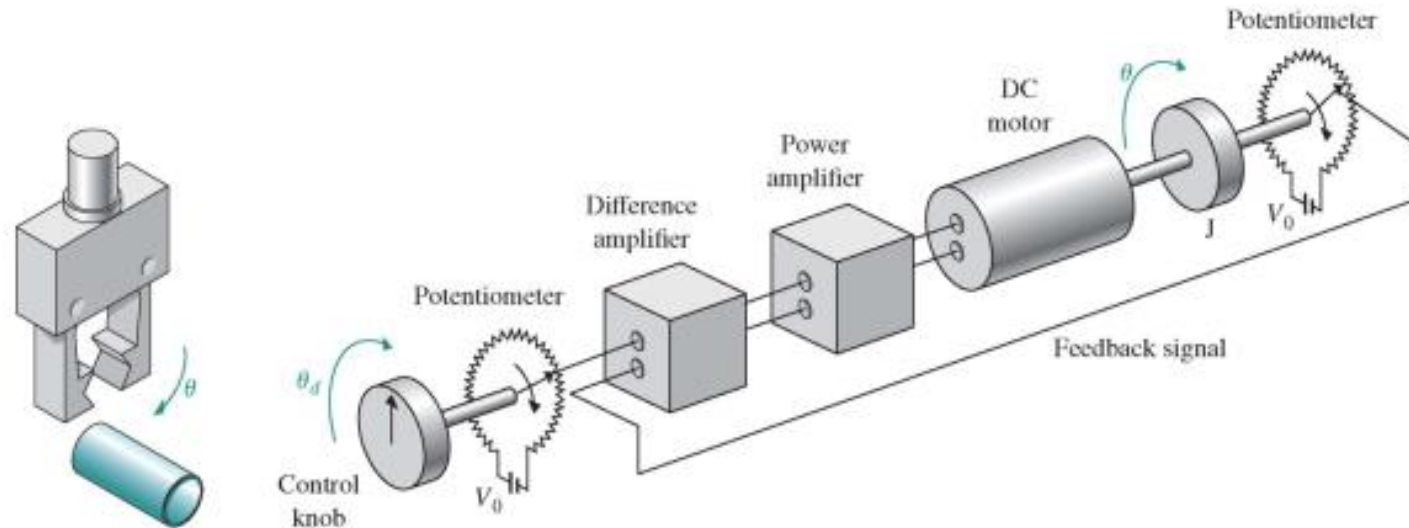
Application

The rotational velocity of the satellite is adjusted by changing the length of the beam. How can we determine the shape of the transient response?



Applications

A robot gripper is to be controlled by a DC motor. How can we determine the transient response of the gripper's position?



From the last lecture

A transfer function can be written as

$$H(s) = \frac{N(s)}{D(s)} = \frac{b_m s^m + b_{m-1} s^{m-1} + \dots + b_1 s + b_0}{a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0} \quad (1)$$

with $n \geq m$. The zeros z_i are the roots of

$$N(s) = 0$$

Thus:

$$\lim_{s \rightarrow z_i} N(s) = 0 \quad (2)$$

The poles p_i are the roots of

$$D(s) = 0$$

Thus:

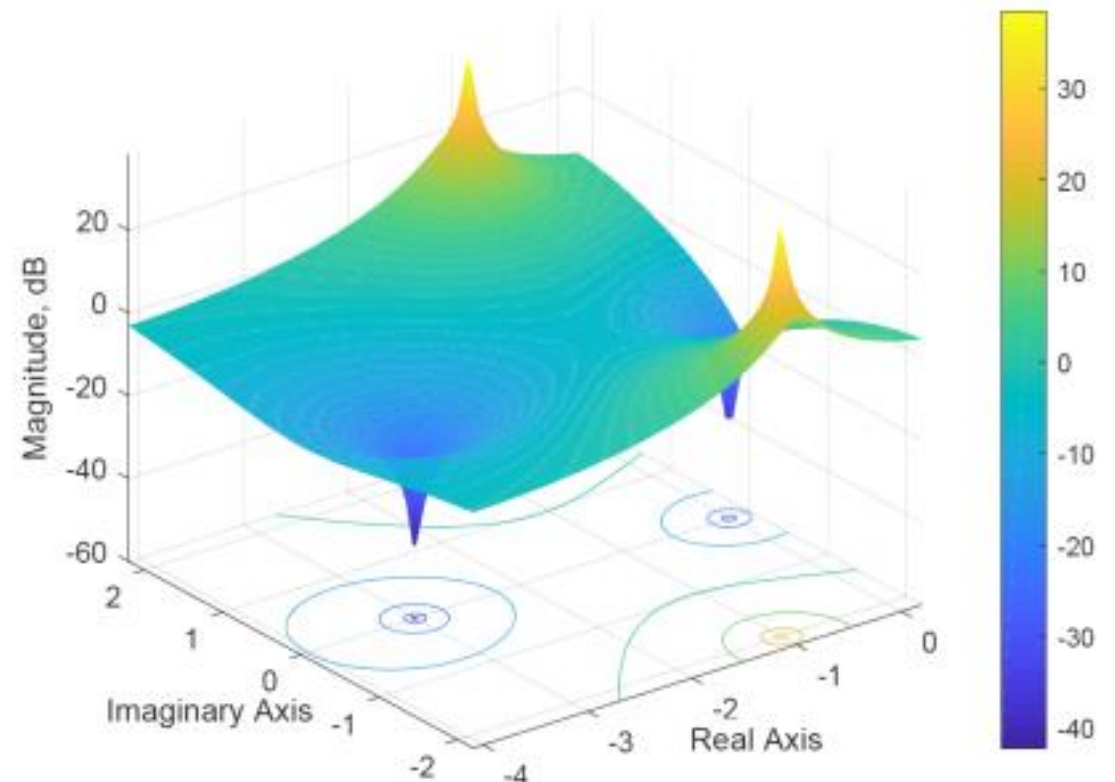
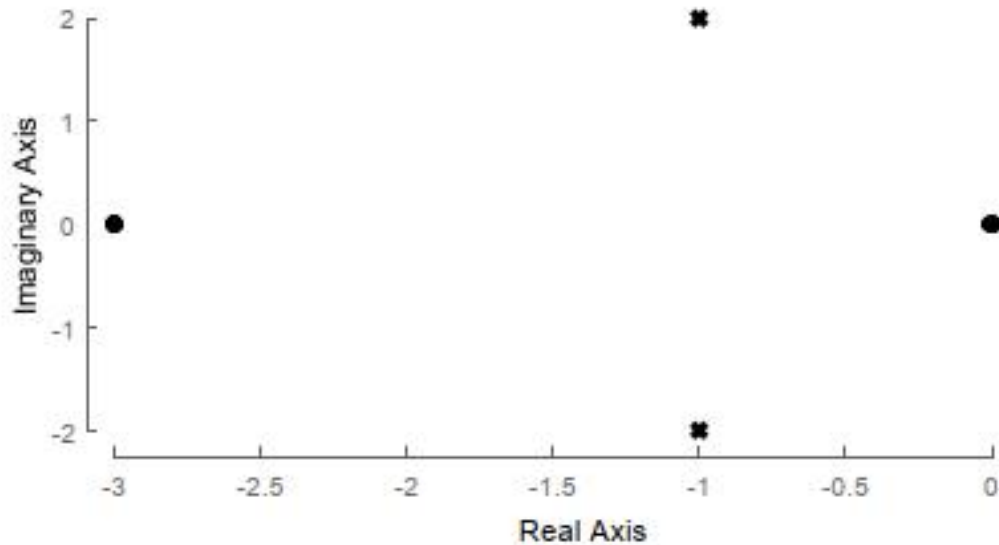
$$\lim_{s \rightarrow p_i} H(s) = \infty \quad (3)$$

Consider the following function:

$$F(s) = \frac{s(s + 3)}{s^2 + 2s + 5}$$

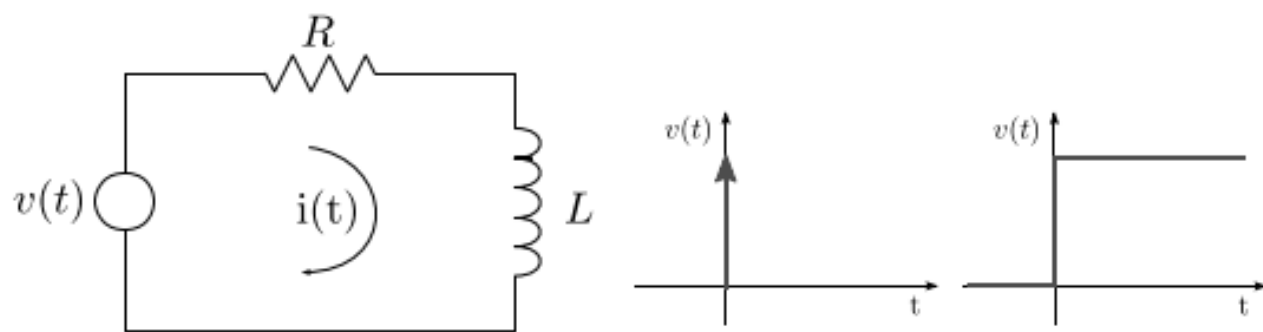
→ Poles: $-1 + 2j$, $-1 - 2j$

→ Zeros: 0 , -3



First order systems

Consider the RL circuit shown.



$$V(s) = (R + Ls)I(s)$$

$$H(s) = \frac{I(s)}{V(s)} = \frac{1}{Ls + R} = \frac{1/R}{s(L/R) + 1}$$

Time constant: Characterizes the response to a step input of a first-order system.

$$\tau = \frac{L}{R} \tag{4}$$

→ The denominator must be in the form of $\tau s + \mathbf{1}$

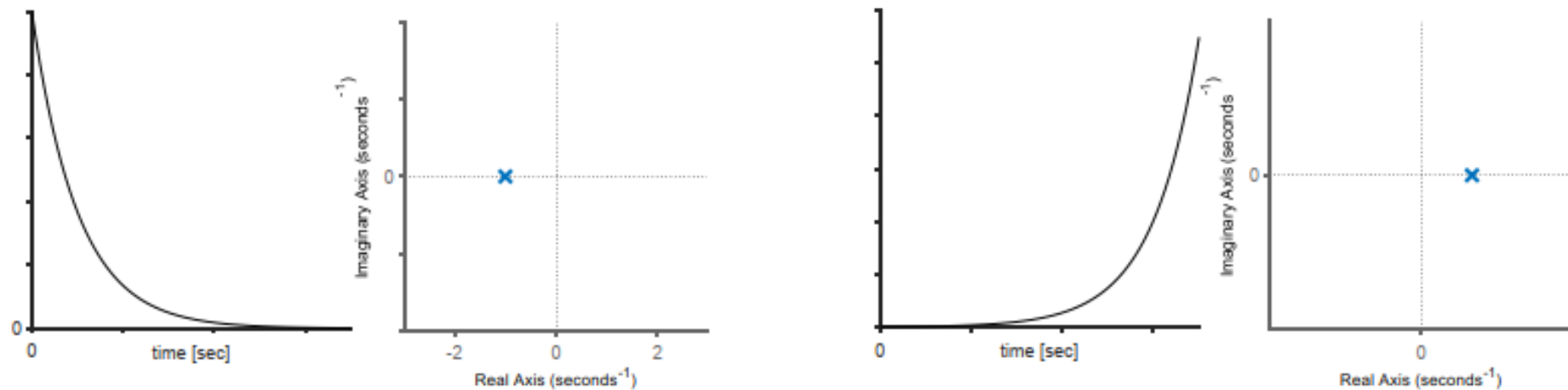
First order transfer functions

Impulse response: $v(t) = \delta(t) \Rightarrow V(s) = 1$

$$I(s) = \frac{1}{R} \frac{1}{\tau s + 1} = \frac{1}{\tau R} \left(\frac{1}{s + \frac{1}{\tau}} \right)$$

The pole is $s = -1/\tau$. The time response is:

$$i(t) = \frac{1}{\tau R} e^{-\frac{t}{\tau}}$$



If $\tau > 0$, the pole is on the left-half s-plane.

If $\tau < 0$, the pole is on the right-half s-plane.

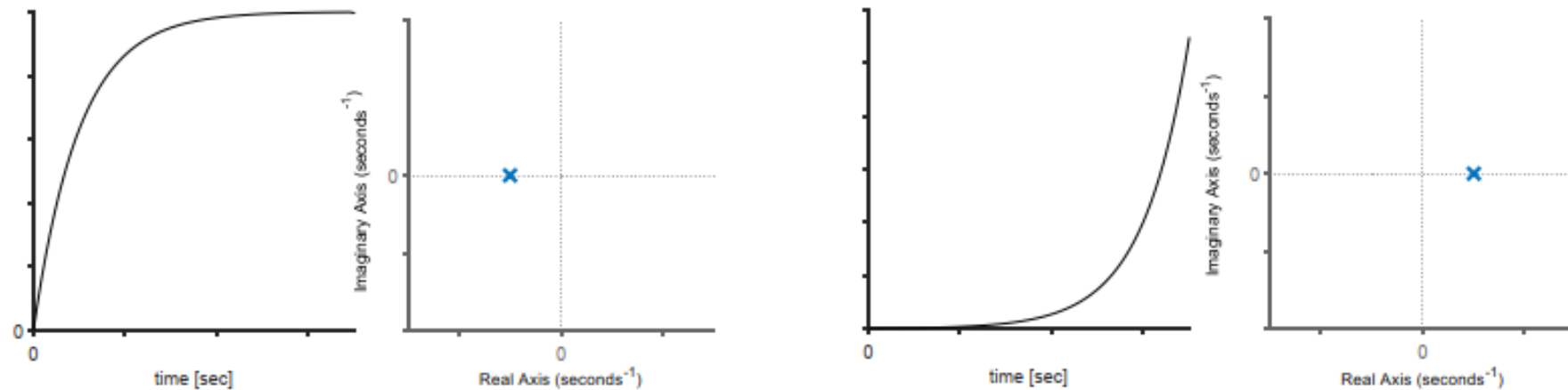
First order transfer functions

Step response: $v(t) = u(t) \Rightarrow V(s) = \frac{1}{s}$

$$I(s) = \frac{1}{\tau R} \left(\frac{1}{s} \frac{1}{s + \frac{1}{\tau}} \right) = \frac{1}{\tau R} \left(\frac{1}{s} \right) \left(\frac{1}{s + \frac{1}{\tau}} \right) = \frac{1}{\tau R} \left(\frac{k_1}{s} + \frac{k_2}{s + \frac{1}{\tau}} \right)$$

Solving for the partial fraction coefficients: $k_1 = \tau$, $k_2 = -\tau$, thus:

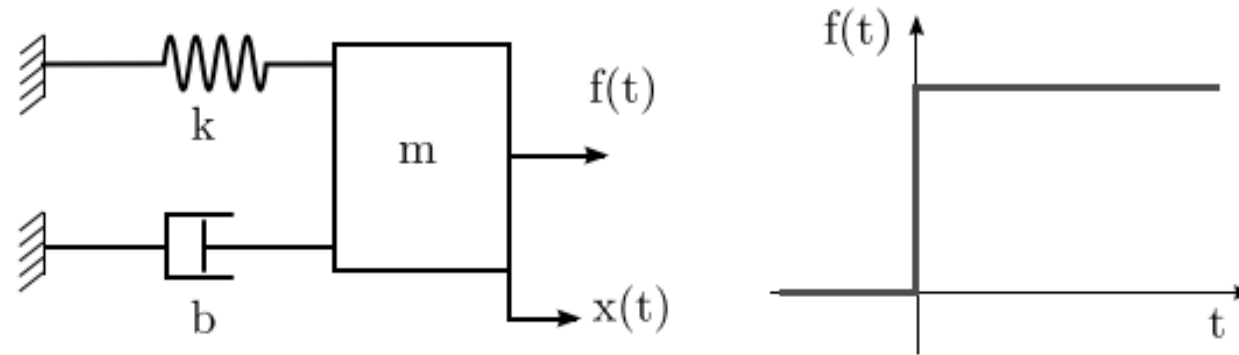
$$i(t) = \frac{1}{R} \left(1 - e^{-\frac{t}{\tau}} \right)$$



If $\tau > 0$, the pole is on the left-half s-plane.

If $\tau < 0$, the pole is on the right-half s-plane.

Second order transfer functions



Transfer function: Standard form

$$H(s) = \frac{X(s)}{F(s)} = \frac{1}{ms^2 + bs + k} = \frac{1}{m} \left(\frac{1}{s^2 + \frac{b}{m}s + \frac{k}{m}} \right)$$

In standard form we have

$$H(s) = \frac{1}{k} \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \quad (5)$$

where:

$$\Rightarrow \zeta = \frac{b}{2\sqrt{mk}}: \text{ Dimensionless **damping ratio** }$$

$$\Rightarrow \omega_n = \sqrt{\frac{k}{m}}: \text{ Natural frequency (rad/s) }$$

Second order response

Let us now analyse the response to a step input of a second order system

$$X(s) = \frac{1}{k} \left(\frac{1}{s} \right) \left(\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \right) \quad (6)$$

The poles of the transfer function are:

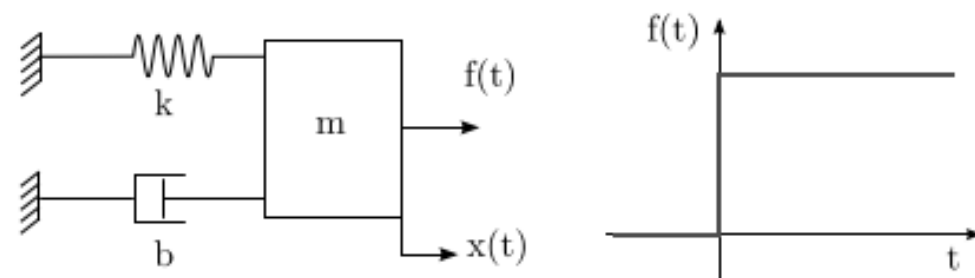
$$s = \frac{-2\zeta\omega_n \pm \sqrt{(2\zeta\omega_n)^2 - 4\omega_n^2}}{2} = -\zeta\omega_n \pm \omega_n\sqrt{\zeta^2 - 1} \quad (7)$$

Thus:

$$s_1 = \omega_n \left(-\zeta + \sqrt{\zeta^2 - 1} \right) = \omega_n \left(-\zeta + j\sqrt{1 - \zeta^2} \right)$$
$$s_2 = \omega_n \left(-\zeta - \sqrt{\zeta^2 - 1} \right) = \omega_n \left(-\zeta - j\sqrt{1 - \zeta^2} \right)$$

Roots can be real or complex \Rightarrow Inverse transformation ?

Second order response



$$X(s) = \frac{1}{k} \left(\frac{1}{s} \right) \left(\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \right) \quad (8)$$

Case 1: $\zeta \geq 1$, (lots of damping)

$$s_1 = \omega_n \left(-\zeta + \sqrt{\zeta^2 - 1} \right)$$

$$s_2 = \omega_n \left(-\zeta - \sqrt{\zeta^2 - 1} \right)$$

Roots are negative real numbers. Partial fraction expansion yields:

$$X(s) = \frac{1}{m} \left(\frac{k_1}{s} + \frac{k_2}{(s + a_1)} + \frac{k_3}{(s + a_2)} \right) \quad (9)$$

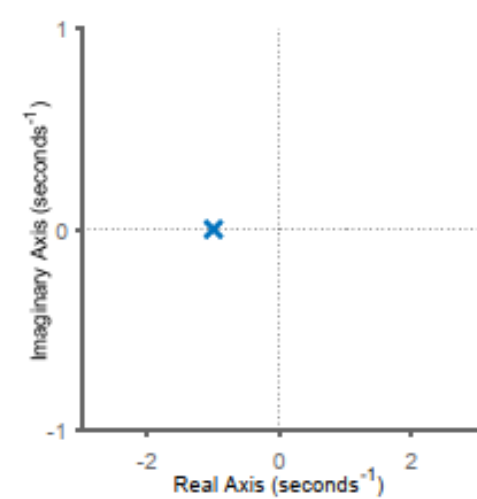
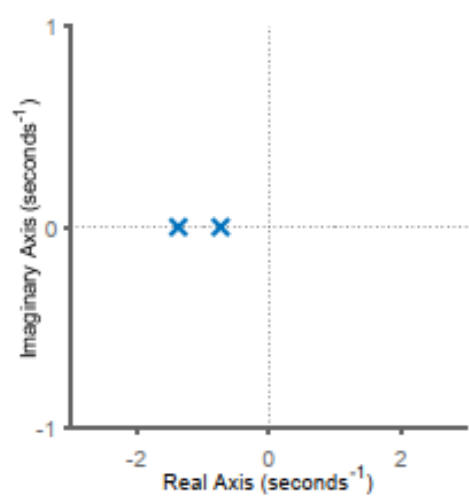
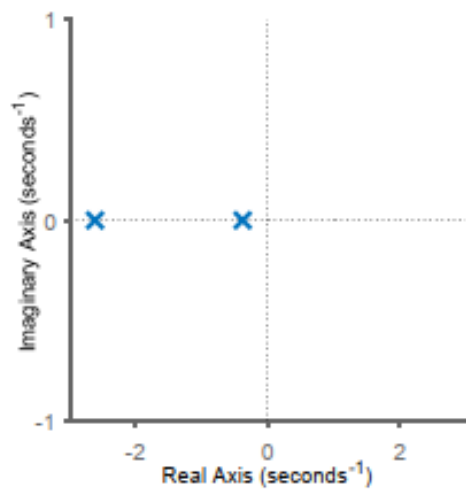
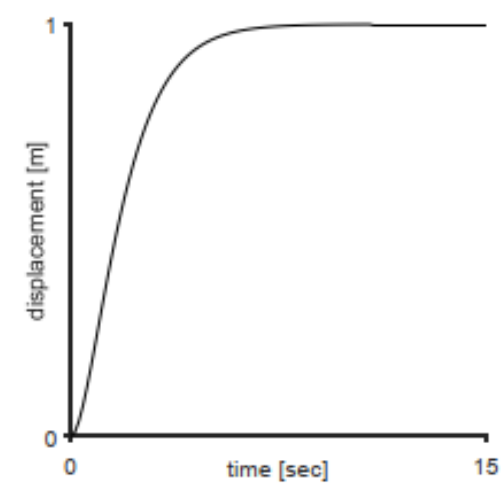
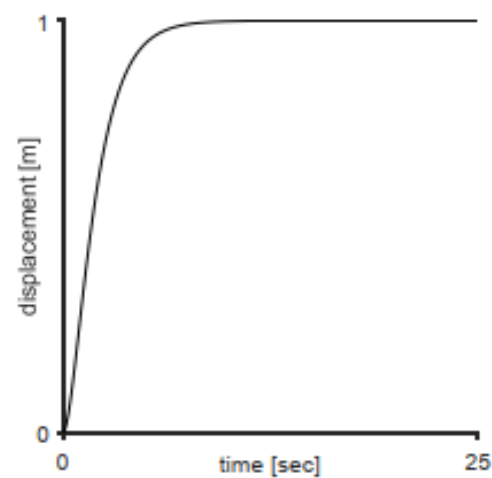
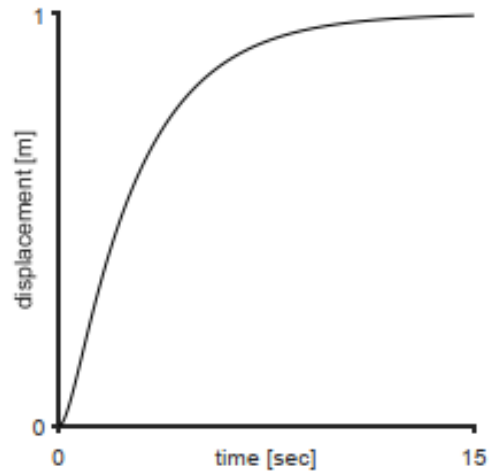
Overdamped system

Example: $m = 1$ kg, $k = 1$ N/m

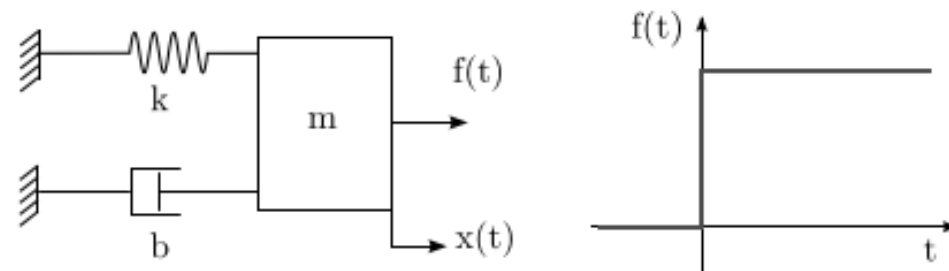
$b = 3$ Ns/m, $\zeta = 1.5$;

$b = 2.1$ Ns/m, $\zeta = 1.05$;

$b = 2$ Ns/m, $\zeta = 1$.



Second order response



$$X(s) = \frac{1}{k} \left(\frac{1}{s} \right) \left(\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \right) \quad (10)$$

Case 2: $0 < \zeta < 1$, (some damping)

$$s_1 = \omega_n \left(-\zeta + \sqrt{\zeta^2 - 1} \right)$$

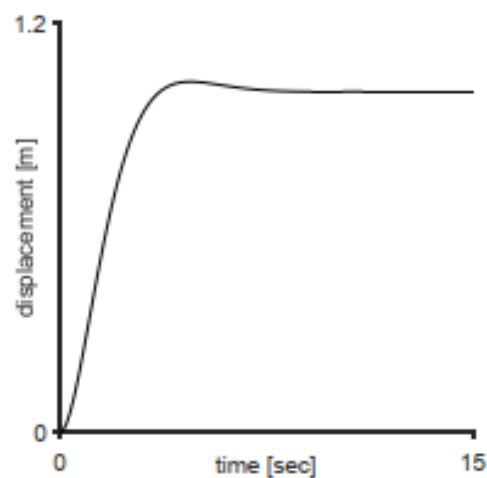
$$s_2 = \omega_n \left(-\zeta - \sqrt{\zeta^2 - 1} \right)$$

Roots are complex conjugate numbers with a negative real part. Thus:

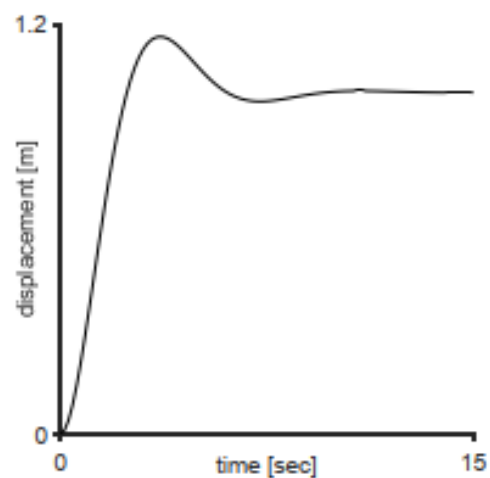
$$x(t) = \frac{1}{k} \left[1 - \frac{1}{\sqrt{1 - \zeta^2}} e^{-\zeta\omega_n t} \sin \left(\omega_n \sqrt{1 - \zeta^2} t + \cos^{-1} \zeta \right) \right] \quad (11)$$

Underdamped system

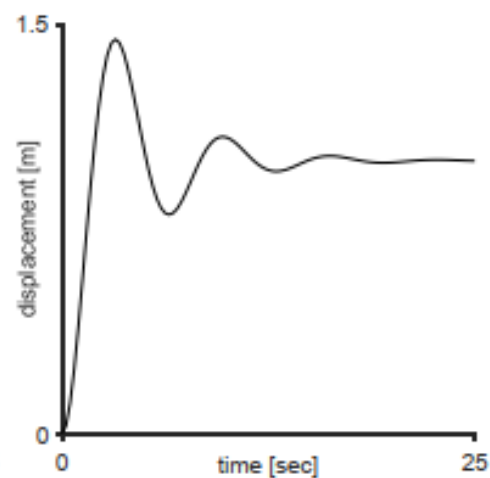
$$b = 1.5, \zeta = 0.75$$



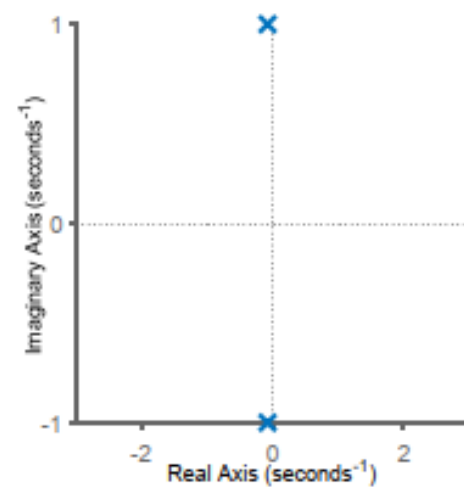
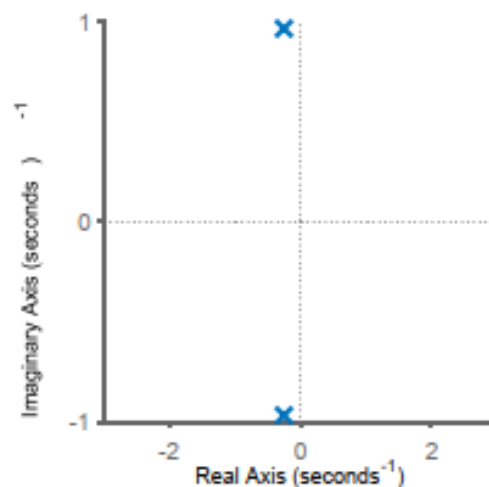
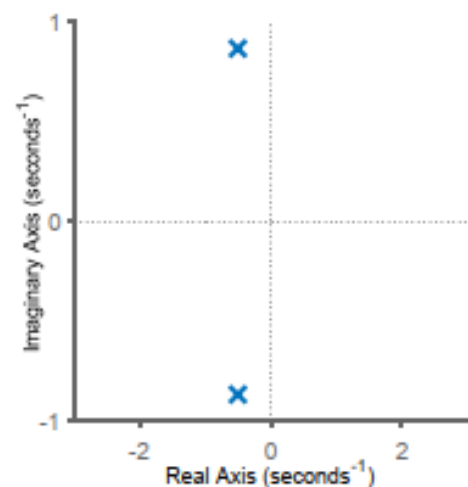
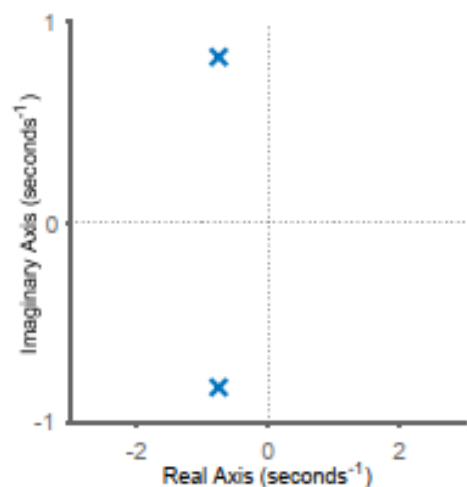
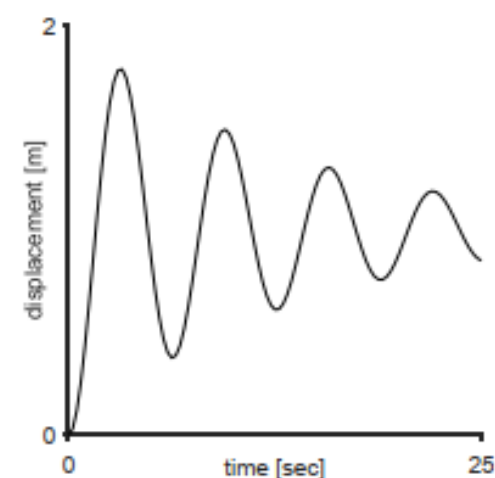
$$b = 1, \zeta = 0.5;$$



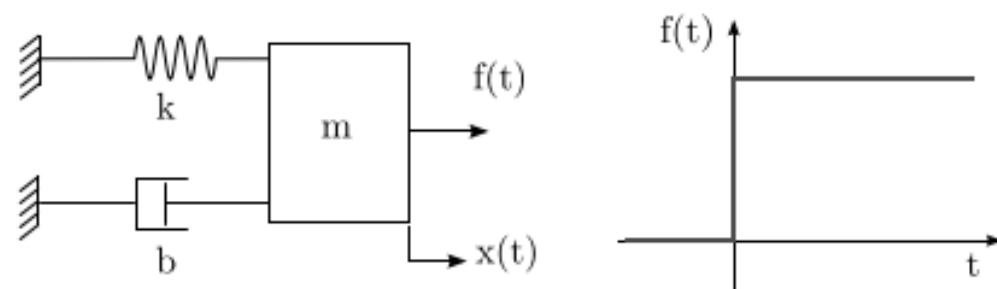
$$b = 0.5, \zeta = 0.25,$$



$$b = 0.1, \zeta = 0.05$$



Second order response



$$X(s) = \frac{1}{k} \left(\frac{1}{s} \right) \left(\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \right) \quad (12)$$

Case 3: $\zeta = 0$, (no damping)

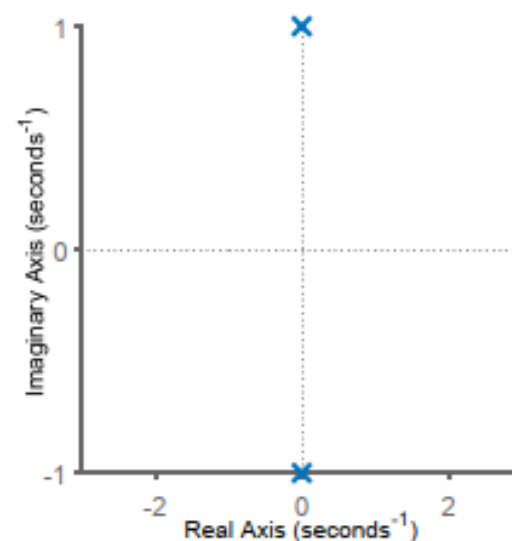
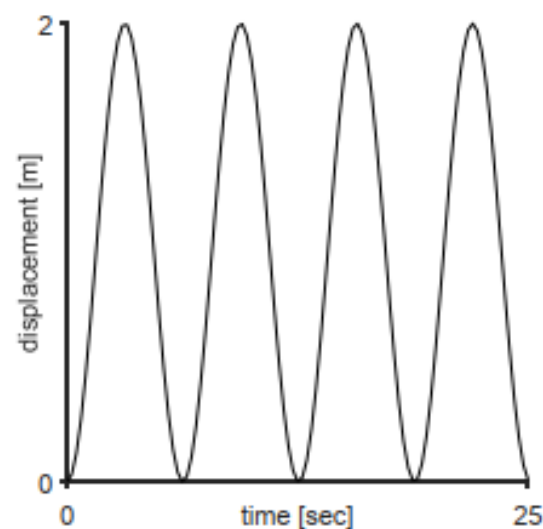
$$s_1 = \omega_n \left(-\zeta + \sqrt{\zeta^2 - 1} \right)$$

$$s_2 = \omega_n \left(-\zeta - \sqrt{\zeta^2 - 1} \right)$$

Roots are purely complex conjugate numbers. Thus:

$$X(s) = \frac{1}{k} \left(\frac{1}{s} \right) \left(\frac{\omega_n^2}{s^2 + \omega_n^2} \right) \rightarrow x(t) = \frac{1}{k} [1 - \cos(\omega_n t)] \quad (13)$$

Undamped system



The frequency of oscillation for of an undamped system is called the natural frequency.

In our example:

$$\omega_n = \sqrt{\frac{k}{m}} \quad (14)$$

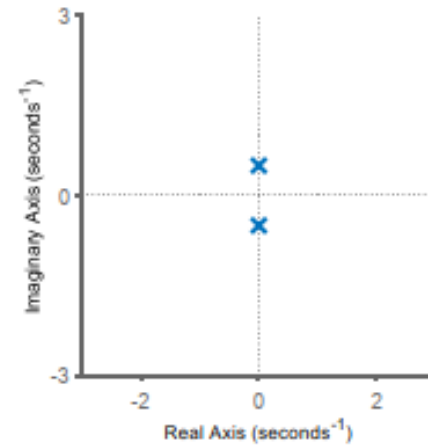
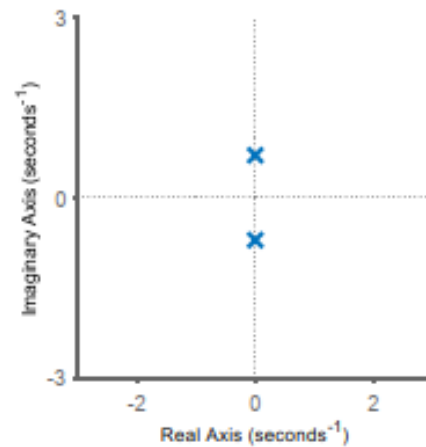
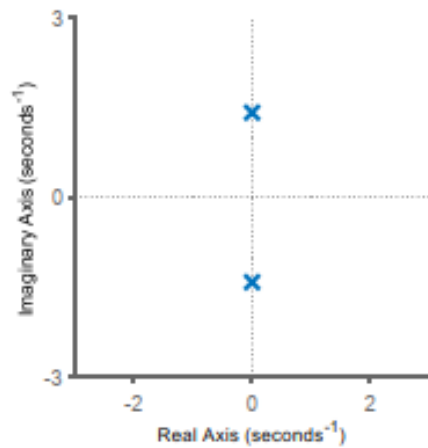
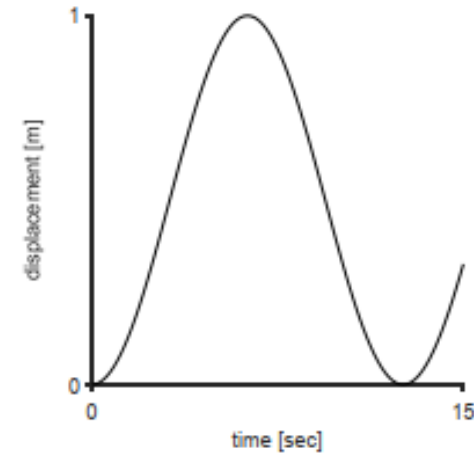
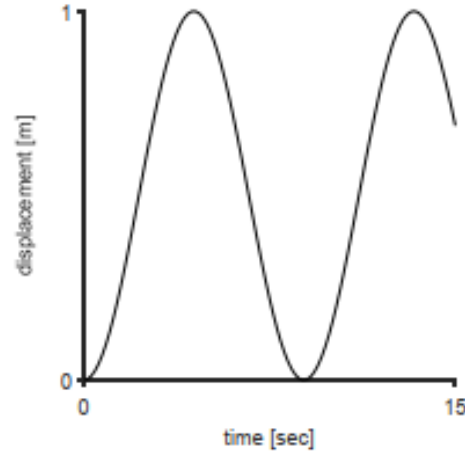
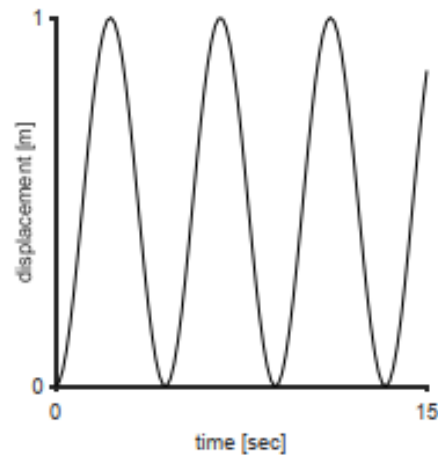
Natural frequency

The frequency the system oscillates when $\zeta = 0$. Example: $b = \zeta = 0, k = 1$.

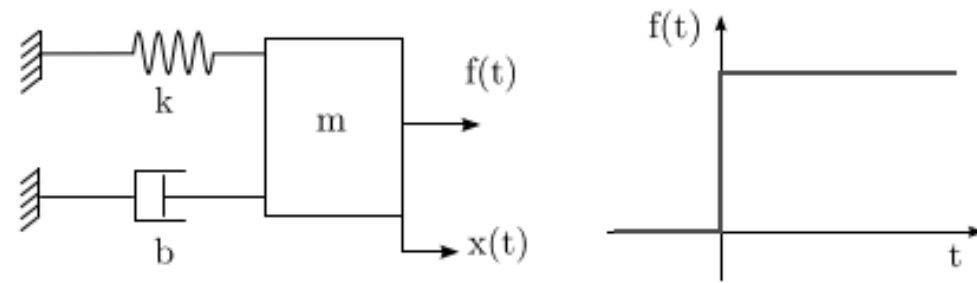
$m = 1, \omega_n = 1 \text{ rad/s};$

$m = 2, \omega_n = 0.71 \text{ rad/s};$

$m = 4, \omega_n = 0.5 \text{ rad/s};$



Second order response



$$X(s) = \frac{1}{k} \left(\frac{1}{s} \right) \left(\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2} \right) \quad (15)$$

Case 4: $\zeta < 0$, (hypothetical negative damping)

$$s_1 = \omega_n \left(-\zeta + \sqrt{\zeta^2 - 1} \right), \quad s_2 = \omega_n \left(-\zeta - \sqrt{\zeta^2 - 1} \right)$$

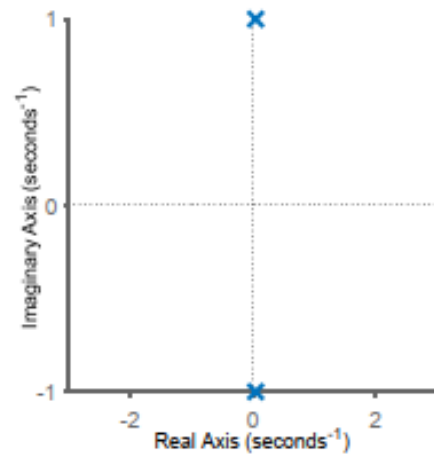
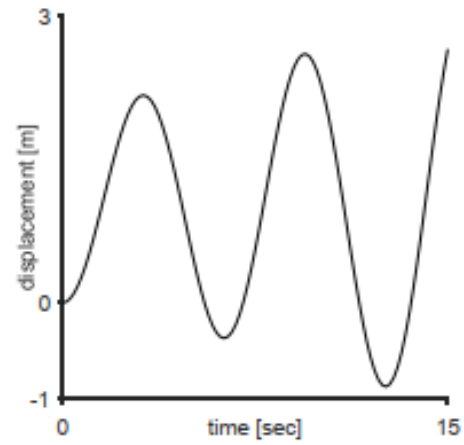
Roots (real or imaginary) have positive real parts. Possible solutions are:

$$x(t) = k(1 + k_2 e^{s_1 t} + k_3 e^{s_2 t}) \quad (16)$$

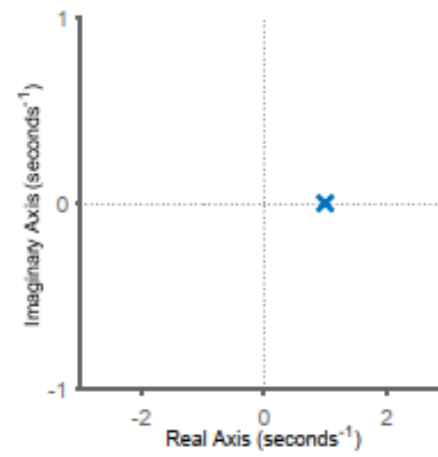
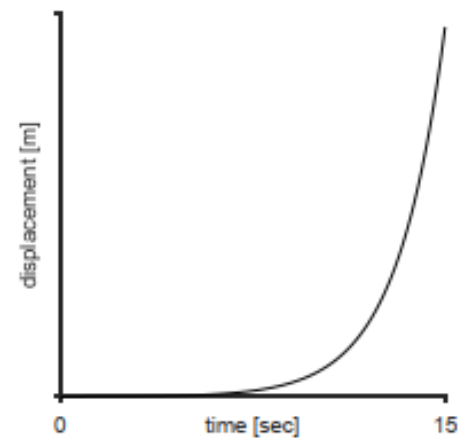
$$x(t) = \frac{1}{k} \left[1 - \frac{1}{\sqrt{1 - \zeta^2}} e^{|\zeta|\omega_n t} \sin \left(\omega_n \sqrt{1 - \zeta^2} t + \cos^{-1} \zeta \right) \right] \quad (17)$$

Unstable system

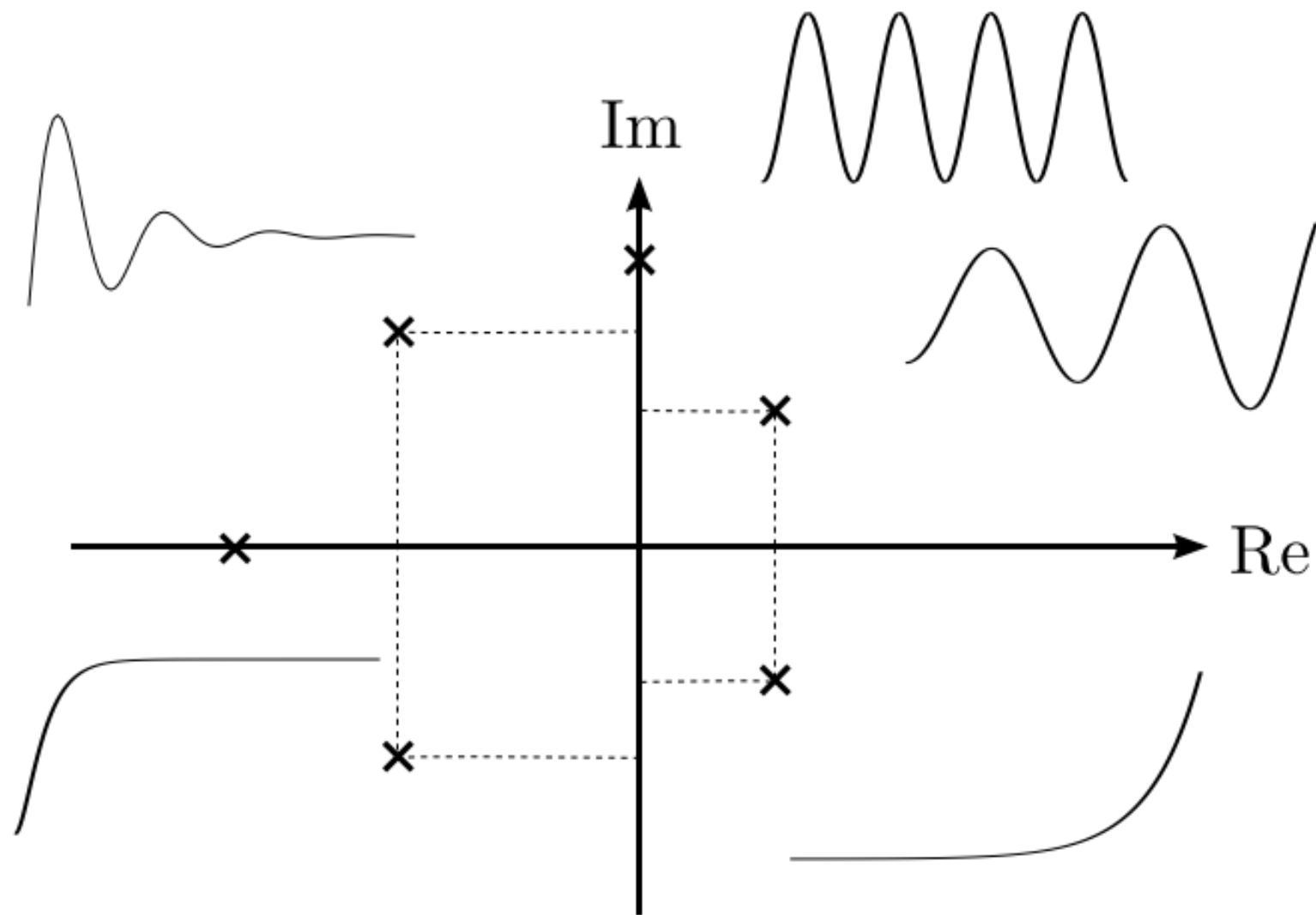
$$b = -0.1 \text{ Ns/m};$$



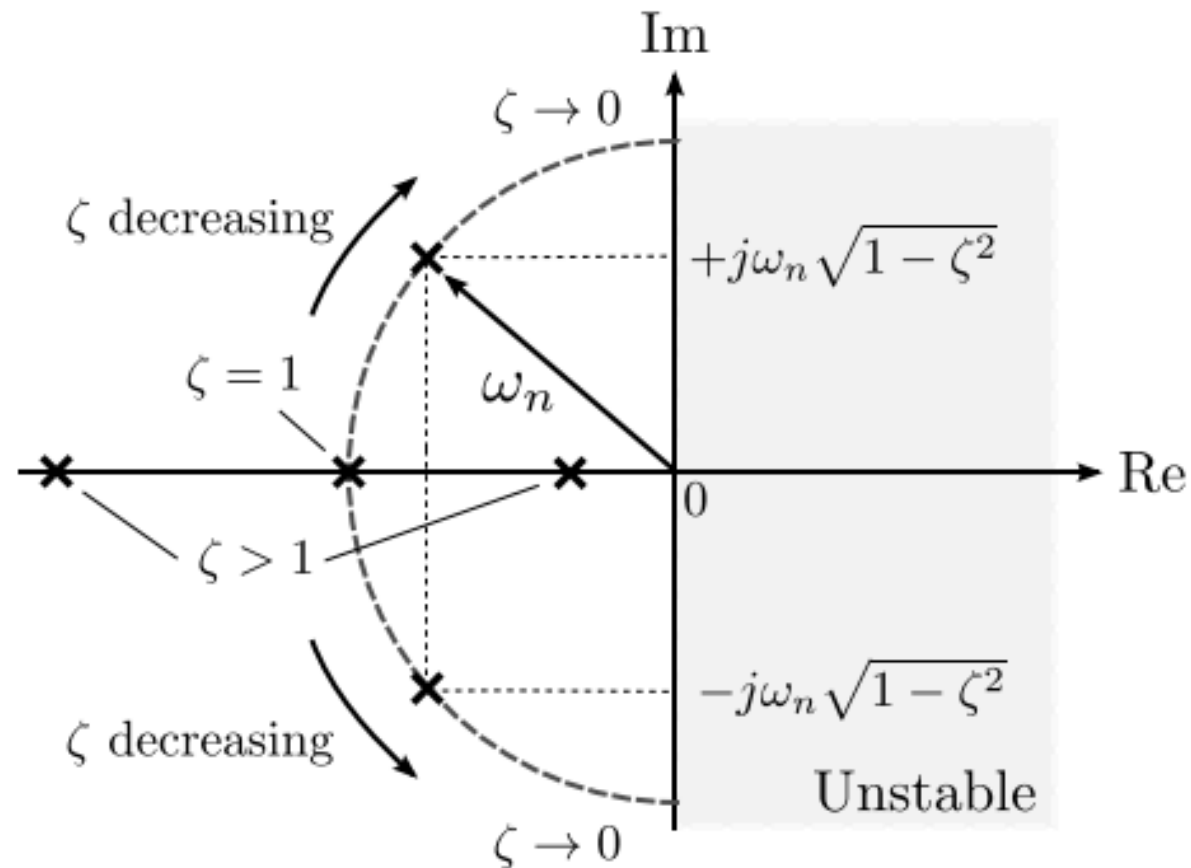
$$b < -1 \text{ Ns/m};$$



Summary



Damping ratio	Roots	Systems response
$\zeta > 1$	Distinct real	overdamped
$\zeta = 1$	Equal real	damped
$0 < \zeta < 1$	Complex conjugate	underdamped
$\zeta = 0$	Purely imaginary	undamped
$\zeta < 0$	Positive	unstable



Location of poles in the s-plane

For a second order system, the poles are

$$s = \zeta\omega_n \pm j\omega_n\sqrt{1 - \zeta^2}$$

$$s = -\sigma \pm j\omega_d$$

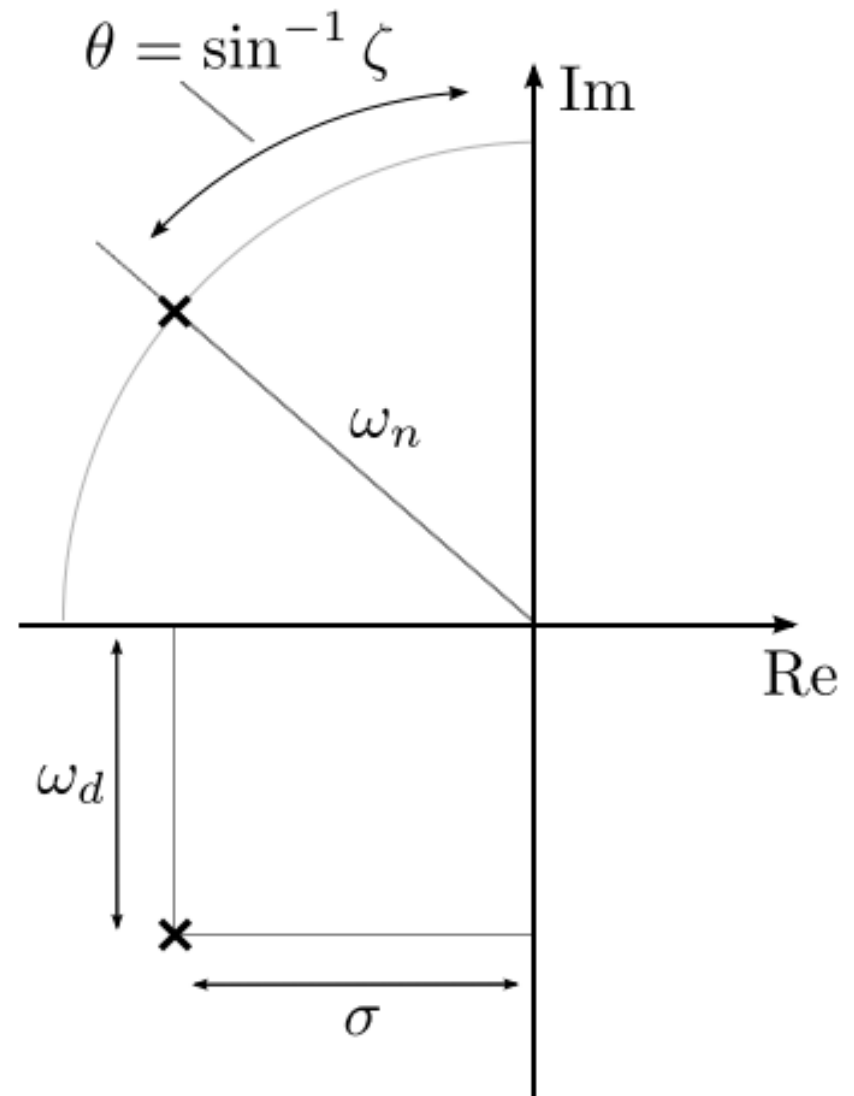
where $\sigma = \zeta\omega_n$, and $\omega_d = \omega_n\sqrt{1 - \zeta^2}$

→ The magnitude of s is

$$|s| = \sqrt{(\zeta\omega_n)^2 + \left(\omega_n\sqrt{1 - \zeta^2}\right)^2} = \omega_n$$

→ The angle to the imaginary axis is

$$\sin \theta = \frac{\zeta\omega_n}{\omega_n} \rightarrow \theta = \sin^{-1} \zeta$$



Exercise 21

Discuss the correlation between the poles of

$$H(s) = \frac{2s + 1}{s^2 + 2s + 5} \quad (18)$$

and the impulse response of the system and find the exact impulse response.

Procedure:

- Calculate the damping ratio and the natural frequency
- Calculate inverse transform

Exercise 21 - continued

$$H(s) = \frac{2s + 1}{s^2 + 2s + 5}$$

Exercise 21 - continued

Exercise 21 - continued

Exercise 22

Consider the system of the form

$$H(s) = \frac{9}{s^2 + bs + 9} \quad (19)$$

Calculate and sketch the response to an unit step input for the following cases

$$\rightarrow b = 9$$

$$\rightarrow b = 0$$

$$\rightarrow b = 2$$

$$\rightarrow b = 6$$

Exercise 22 - continued

$$H(s) = \frac{9}{s^2 + 9s + 9} \frac{1}{s} \quad (20)$$

Exercise 22 - continued

$$H(s) = \frac{9}{s^2 + 9} \frac{1}{s} \quad (21)$$

Exercise 22 - continued

$$H(s) = \frac{9}{s^2 + 2s + 9} \frac{1}{s} \quad (22)$$

Exercise 22 - continued

$$H(s) = \frac{9}{s^2 + 6s + 9} \frac{1}{s} \quad (23)$$

Matlab/Simulink